# Electrical Engineering 229A Lecture 19 Notes

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## 1 Capacity of an Additive White Gaussian Noise Channel

#### 1.1 Shannon capacity of a additive white Gaussian noise channel

In the additive white Gaussian noise (AWGN) model, we send inputs real-valued  $X_i$  and receive real-valued outputs  $Y_i$ , where  $Y_i = X_i + Z_i$ , and  $Z_1, Z_2, \ldots \stackrel{\text{iid}}{\sim} \mathcal{N}(0, \sigma^2)$ . At block length n, we have an encoding map  $e_n : [M_n] \to \mathbb{R}^n$  and a decoding map  $d_n : \mathbb{R}^n \to [M_n]$ . We assume an **input power constraint**, which, in Shannon's formulation, says that each codeword is required to have power at most P: If  $X^n(m)$  denotes  $e_n(m)$ , then

$$\frac{1}{n}\sum_{i=1}^{n}X_{i}^{n}(m)\leq P.$$

We want to find for which rates R we have

$$\liminf_{n} \frac{1}{n} \log M_n \ge R \quad \text{with} \quad \mathbb{P}(d_n(e_n(W_n)) \neq W_n) \to 0$$

for some sequence  $((e_n, d_n), n \ge 1)$  with  $W_n \sim \text{Unif}([M_n])$ .

**Theorem 1.1.** The supremum over rates for which communication is possible is

$$\sup_{X \sim f: \int_{-\infty}^{\infty} x^2 f(x) \, dx \le P} I(X; X + Z),$$

which equals  $\frac{1}{2}\log(1+\frac{P}{\sigma^2})$  and is achieved by  $X \sim N(0, P)$ .

This quantity is called the **Shannon capacity**. The achievability part of the proof will use a random coding argument and requires the concept of  $\varepsilon$ -weakly typical sequences. The converse part of the proof involves Fano's inequality. Let's first see why the last claim is true:

**Lemma 1.1.** If  $\mathbb{E}[X^2] \leq P$ , then  $I(X; X + Z) \leq \frac{1}{2} \log(1 + \frac{P}{\sigma^2})$ , with equality if and only if  $X \sim N(0, P)$ .

Proof.

$$I(X; X + Z) = h(X + Z) - h(X + Z | X)$$
  
=  $h(X + Z) - h(Z)$   
=  $h(X + Z) - \frac{1}{2}\log(2\pi e\sigma^2).$ 

Since  $X \amalg Z$  and  $\mathbb{E}[Z_1] = 0$ , we also have

$$\mathbb{E}[(X+Z)^2] = \mathbb{E}[X^2] + \mathbb{E}[Z^2]$$
$$\leq P + \sigma^2.$$

 $\operatorname{So}$ 

$$h(X+Z) \le \frac{1}{2}\log(2\pi e(P+\sigma^2))$$

with equality iff  $X \sim N(0, P)$ . So

$$I(X, X + Z) \leq \frac{1}{2} \log \left( \frac{P + \sigma^2}{\sigma^2} \right)$$
$$= \frac{1}{2} \log \left( 1 + \frac{P}{\sigma^2} \right).$$

## 1.2 Weak-typicality for differential entropy

**Definition 1.1.** For  $X \sim f$  with differential entropy h(X) and  $\varepsilon > 0$ , the set of  $\varepsilon$ -weakly typical sequences for the density f sis

$$A_{\varepsilon}^{n} := \left\{ x^{n} \in \mathbb{R}^{n} : \left| -\frac{1}{n} \log \prod_{i=1}^{n} f(x_{i}) - h(X) \right| < \varepsilon \right\} \subseteq \mathbb{R}^{n}$$

By the weak law of large numbers,

$$\mathbb{P}(X^n \in A^n_{\varepsilon}) = 1$$

if  $X_i \stackrel{\text{iid}}{\sim} f$ . This is because  $\mathbb{E}[\log \frac{1}{f(X)}] = h(X)$  when  $X \sim f$ . **Proposition 1.1.** For all n,

$$\operatorname{Vol}(A_{\varepsilon}^n) \le 2^{nh(X)} 2^{n\varepsilon}.$$

Proof.

$$1 \ge \int_{A_{\varepsilon}^{n}} \prod_{i=1}^{n} f(x_{i}) dx^{n}$$
  
$$\ge \int_{A_{\varepsilon}^{n}} 2^{-nh(X)} 2^{-n\varepsilon} dx^{n}$$
  
$$= \operatorname{Vol}(A_{\varepsilon}^{n}) 2^{-nh(X)} 2^{-n\varepsilon}.$$

**Proposition 1.2.** Given  $\delta > 0$ , for all sufficiently large n,

$$\operatorname{Vol}(A_{\varepsilon}^n) \ge (1-\delta)2^{nh(X)}2^{-n\varepsilon}$$

*Proof.* For sufficiently large n,

$$(1 - \delta) \leq \int_{A_{\varepsilon}^{n}} \prod_{i=1}^{n} f(x_{i}) dx^{n}$$
$$\leq \int_{A_{\varepsilon}^{n}} 2^{-nh(X)} 2^{n\varepsilon} dx^{n}$$
$$= \operatorname{Vol}(A_{\varepsilon}^{n}) 2^{-nh(X)} 2^{n\varepsilon}.$$

**Definition 1.2.** Let  $(X_1, Y_1), (X_2, Y_2), \ldots$  be iid with  $(X_i, Y_i) \sim f(x, y)$ . The set of  $\varepsilon$ jointly weakly typical sequences for f is

$$A_{\varepsilon}^{n} := \left\{ (x^{n}, y^{n}) : \left| -\frac{1}{n} \log \prod_{i=1}^{n} f(x_{i}) - h(X) \right| \le \varepsilon, \\ \left| -\frac{1}{n} \log \prod_{i=1}^{n} f(y_{i}) - h(Y) \right| \le \varepsilon, \\ \left| -\frac{1}{n} \log \prod_{i=1}^{n} f(x_{i}, y_{i}) - h(X, Y) \right| \le \varepsilon, \right\}.$$

With this definition in mind, we can show the following.

**Lemma 1.2.** If  $\widetilde{X}^n \stackrel{d}{=} X^n$ ,  $\widetilde{Y}^n \stackrel{d}{=} Y^n$ , and  $\widetilde{X}^n \amalg \widetilde{Y}^n$ , then

$$(1-\delta)2^{-nI(X;Y)}2^{-3n\varepsilon} \le \mathbb{P}((\widetilde{X}^n, \widetilde{Y}^n) \in A_{\varepsilon}^n) \le 2^{-nI(X;Y)}2^{3n\varepsilon}.$$

The upper bound holds for all n, and the lower bound holds for all sufficiently large n.

**1.3** Proof of Shannon's channel coding theorem for an AWGN channel Now we can prove the theorem.

Proof. Achievability: Generate a random codebook

$$\begin{bmatrix} X_1(1) & \cdots & X_n(1) \\ X_1(2) & \cdots & X_n(2) \\ \vdots & & \vdots \\ X_1(M_n) & \cdots & X_n(M_n) \end{bmatrix},$$

where each  $X_n(i) \sim \mathcal{N}(0, P - \eta)$  is iid over *i* and *n*. Let  $W_n \sim \text{Unif}([M_n])$ . The decoding rule is

$$d_n(Y^n) = \begin{cases} m & (X^n(m), Y^n) \text{ are } \varepsilon\text{-jointly weakly typical and for all } m' \neq m \\ & (X^n(m), Y^n) \text{ are not } \varepsilon\text{-jointly weakly typical} \\ \text{arbritrary} & \text{either no or } \geq 2 \ X^n(m) \text{ are } \varepsilon\text{-jointly typical with } Y^n. \end{cases}$$

By symmetry,

$$\mathbb{P}(d_n(e_n(W_n)) \neq W_n) = \mathbb{P}(d_n(e_n(1) \neq 1))$$
  
$$\leq P(E_{0,n}) + \sum_{m \neq 2} P(E_{m,n}),$$

where  $E_{0,n}$  is the event that  $(X^n(1), Y^n)$  is not  $\varepsilon$ -jointly weakly typical and  $E_{m,n}$  for  $m \ge 2$ is the event that  $(X^n(1), Y^n)$  is  $\varepsilon$ -jointly weakly typical. Then  $\mathbb{P}(E_{0,n}) \to 0$  as  $n \to \infty$ , and for each  $2 \le m \le M_n$ ,  $\mathbb{P}(E_{m,n}) \le 2^{-nI(X;Y)}2^{3n\varepsilon}$ . So if  $M_n = 2^{nR}$  with  $R < I(X;Y) - 3\varepsilon$ , then  $\mathbb{P}(d_n(e_n(W_n)) \ne W_n) \to 0$  as  $n \to \infty$ .

Converse: Consider any  $((e_n, d_n), n \ge 1)$ . We have  $W_n \sim \text{Unif}([M_n])$  and the Markov chain  $W_n - X^n - Y^n - \widehat{W}_n$  with  $X^n = e_n(W_n)$ , Y = X + Z, and  $\widehat{W}_n = d_n(Y^n)$ . Suppose  $M_n = \lceil 2^{nR} \rceil$ . The data-processing inequality gives

$$H(W_n \mid Y^n) \le H(W_n \mid \widehat{W}_n).$$

Note that  $W_n$  is a discrete random variable, and  $Y^n$  is a continuous random variable. Here, we mean  $H(W_n | Y^n) = \int_{-\infty}^{\infty} H(W_n | Y_n = y) f(y) \, dy$ . If  $p_e(n) := \mathbb{P}(\widehat{W}_n \neq W_n)$ , then Fano's inequality gives

$$H(W_n \mid Y^n) \le 1 + nRp_e(n).$$

Also, the data processing inequality gives

$$H(W_n) = I(W_n; Y^n) + H(W_n \mid Y^n)$$
  

$$\leq I(X^n; Y^n) + H(W_n \mid Y^n)$$
  

$$= h(Y^n) - \sum_{i=1}^n h(Y_i \mid X^n, Y^{i-1}) + H(W_n \mid Y^n)$$

Use  $0 \le D(f(y^n) || \prod_{i=1}^n f(y_i)) = \int_{\mathbb{R}^n} f(y^n) \log \frac{f(y^n)}{\prod_{i=1}^n f(y_i)} dy^n = -h(Y^n) + \sum_{i=1}^n h(Y_i).$ 

$$\leq \sum_{i=1}^{n} h(Y_i) - \sum_{i=1}^{n} h(Y_i \mid X^n, Y^{i-1}) + H(W_n \mid Y^n)$$

Use the Markov chain  $Y_i - X_i - (X^{i-1}, X^n_{i+1}, Y^{i-1})$ n

$$\leq \sum_{i=1}^{n} h(Y_i) - \sum_{i=1}^{n} h(Y_i \mid X_i) + H(W_n \mid Y^n)$$

$$= \sum_{i=1}^{n} I(X_{i}; Y_{i}) + H(W_{n} \mid Y^{n})$$

Let  $P_i := \mathbb{E}[X_i^2]$ , and recall that  $Y_i = X_i + Z_i$ , where  $Z \sim \mathcal{N}(0, \sigma^2)$  and  $X_i \amalg Z_i$ .

$$\leq \sum_{i=1}^{n} \frac{1}{2} \log \left( 1 + \frac{P_i}{\sigma^2} \right) + H(W_n \mid Y^n)$$
  
$$\leq \frac{n}{2} \log \left( 1 + \frac{P}{\sigma^2} \right) + H(W_n \mid Y^n)$$
  
$$\leq \frac{n}{2} \log \left( 1 + \frac{P}{\sigma^2} \right) + 1 + (\log M_n) p_e(n).$$

Since  $\frac{1}{n} \log M_n \to R$  if  $p_e(n) \to 0$ , this gives

$$\limsup_{n} \frac{1}{n} \log M_n \le \frac{1}{2} \log \left( 1 + \frac{P}{\sigma^2} \right). \qquad \Box$$

Why is this result interesting? Suppose the FCC assigns you a bandwidth of W Hertz, and you communicate over this channel for some time T at power constraint P (with units energy per unit time). One can show that if the noise that corrupts your waveform is additive white noise with power spectral density  $\frac{N_0}{2}$ , then the theoretical limit of the rate at which you can communicate is

$$W \log(1 + \frac{P}{N_0 W})$$
 bits/unit time.

Studying the  $W \to \infty$  limit and the  $T \to \infty$  limit is interesting.